

# An improved mathematical modeling to simulate metallization screen pattern trend for silicon solar cell

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**Abstract** — the series resistance calculation based on H-type model has been widely used to understand power loss mechanism in semiconductor solar cell, and one of its popular applications is to predict optimal metallization screen printing pattern. The disadvantage of this model is that it is not accurate to estimate the relationship of screen pattern trend vs. cell efficiency, since it is based on an ideal rectangular finger shape, which does not exist in present regular screen printing process. In this work, a revised model is proposed to simulate the contact resistant power loss in multi-crystalline silicon solar cell. Improved accuracy of the new model is found by comparing the simulation results to the I-V testing data of multi-crystalline silicon solar cell printed with Heraeus 96XX front Ag paste.

**Index Terms** — silicon, solar cell, series resistance, metallization

## I. INTRODUCTION

As the dominate solar-cell metallization technique in present silicon solar cell manufacturing, screen printing has superior cost saving over other technologies. In the past decade, paste suppliers have been responsible for steady improvements in cell performance and reduction in dollar-per-watt costs, and nowadays are facing more challenges to further cost reduction and to improve cell efficiency. Therefore, a more accurate understanding of optimal grid pattern trend vs. cell efficiency is needed.

The popular H-type grid pattern (Fig.1) for series resistance simulation is based on the assumptions [1]: (i) the cell is uniformly illuminated; (ii) the photo current flows uniformly into the surface sheet and spreads laterally to the gridline and then collected by bus bar; (iii) grid line fingers and bus bar cross section area are rectangular; (iv) fingers are uniform; (v) finger width is wider than transfer lifetime  $L_T$  by 1.5 times.

However, in practice screen printed fingers seldom hold ideal rectangular shape, and are usually not uniform. In addition, the finger width in present cell manufacture processes are significantly reduced, from 100–200  $\mu\text{m}$  three years ago, to no longer wider than 1.5 times of  $L_T$  today. These differences will cause discrepancies between I-V tested series resistance values and calculated values, hence, lead to inaccurate cell performance estimates. In this work, a revised model is proposed to simulate the series resistant power loss in multi-

crystalline silicon solar cell, greatly reduced the difference between the calculation and tested results.

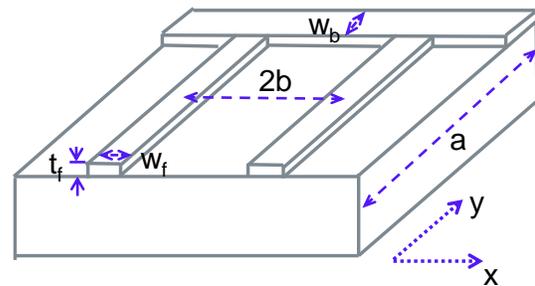


Fig.1 H-type grid pattern for rectangular solar cell

## II. REVISED MODEL DESCRIPTION

In the revised model, we begin at the unit cell with  $a$  units long and  $2nb$  units wide, with  $n$  the number of grid fingers, and  $2b$  the distance between the centers of neighboring grid fingers. The entire rectangular solar cell is then composed of repeated unit cells. The power loss due to series resistance is associated with current travelling through the emitter, the contact resistance, gridline fingers, bus bars, and through the base of cells. For diffused emitter layer and base, D. Meier gave the expressions as below [1],

$$P_{sheet} = \frac{2}{3} J_L^2 nab^3 R_{sheet} \quad (1)$$

$$P_{base} = 2J_L^2 nabl\rho_{base} \quad (2)$$

Where  $J_L$  is the light-generated current density,  $R_{sheet}$  is the emitter sheet resistance,  $l$  is the base region thickness, and  $\rho_{base}$  is the semiconductor base resistivity.

Since the screen printed grid finger shape are not ideally rectangular, we revise grid finger cross section area by a trapezoid shape or Gaussian shape to simulate the practical screen printing finger shape, as shown in Fig.2-(a)&(b).

The element resistance  $dR$  of the grid finger is given by,

$$dR = (\rho_f/A) dy \quad (3)$$

Where  $A$  is the cross-sectional area of the grid finger, and  $\rho_f$  is the gridline resistivity.

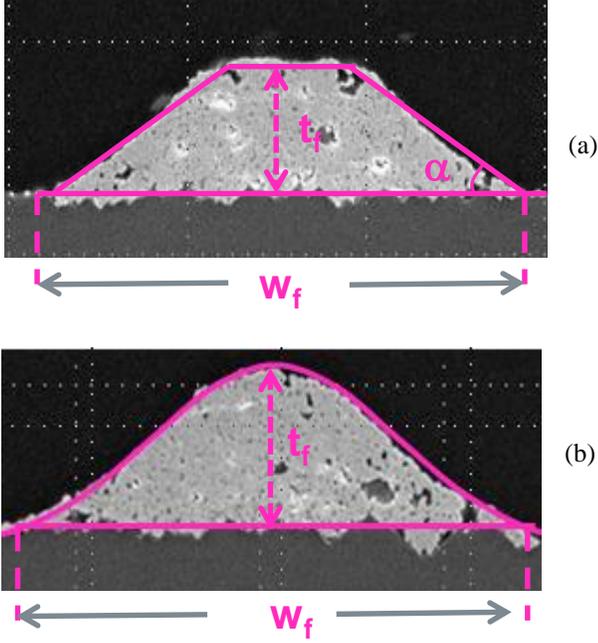


Fig.2 Screen printing grid finger cross-section area:  
(a) Trapezoid shape, (b) Gaussian shape

For a trapezoid grid finger as shown in Fig.2-(a), cross-section area is,

$$A_1 = (W_f - \frac{t_f}{tg\alpha})t_f \quad (4)$$

Where  $W_f$  is the finger width,  $t_f$  is the finger height, and  $tg\alpha$  is the slope of finger side-wall. For a given  $tg\alpha$  and  $W_f$ , the finger height range is  $0 < t_f < 0.5W_f tg\alpha$ .

For a Gaussian shape grid finger like Fig.2-(b), the profile of finger can be described as equation (5), with the peak of  $t_f$  and the standard deviation of  $\sigma$ ,

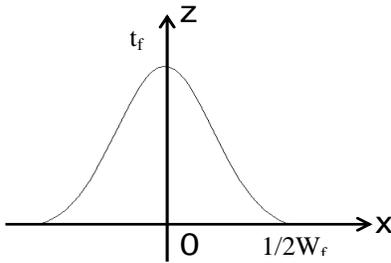


Fig.3 Gaussian profile of grid finger

$$\varphi(x) = t_f e^{-\frac{x^2}{2\sigma^2}} \quad (5)$$

The area within Gaussian profile can be obtained in equation (6), and the full width at half maximum (FWHM) is related to the standard deviation  $\sigma$  according to equation (7),

$$A_2 = \int_{-\infty}^{+\infty} \varphi(x) dx = t_f \sigma \sqrt{2\pi} \quad (6)$$

$$FWHM = 2\sqrt{2\ln 2} \sigma \quad (7)$$

If we define  $s = FWHM/W_f$ , the cross-section area yields,

$$A_2 = \frac{sw_f t_f \sqrt{\pi}}{2\sqrt{\ln 2}} \quad (8)$$

Substitute equation (4) and (8) into equation (3), the element resistance of trapezoid and Gaussian shape grid fingers are given in equation (9) and (10), respectively,

$$dR1 = \frac{\rho_f}{(W_f - \frac{t_f}{tg\alpha})t_f} dy \quad (9)$$

$$dR2 = (\frac{2\rho_f \sqrt{\ln 2}}{sw_f t_f \sqrt{\pi}}) dy \quad (10)$$

Since the power loss associated with current flow along the grid finger is (11),

$$P_{finger} = n \int_0^a I^2(y) dR \quad (11)$$

$$I(y) = 2b(a - y)J_L \quad (12)$$

The power loss of grid fingers with trapezoid or Gaussian shape cross-section areas can be written as (13) and (14), respectively,

$$P_{finger1} = \frac{4}{3} J_L^2 n a^3 b^2 \rho_f \frac{1}{(w_f - \frac{t_f}{tg\alpha})t_f} \quad (13)$$

$$P_{finger2} = \frac{8}{3} J_L^2 n a^3 b^2 \rho_f \frac{\sqrt{\ln 2}}{s w_f t \sqrt{\pi}} \quad (14)$$

Now we consider another difference between practical grid fingers to the ideal ones, uniformity. As shown in Fig. 4, the roughness of the real gridline fingers can cause a disagreement in line resistance to the ideal smooth case. Since the rough surface is mainly formed by the wire mesh of the screen, the peaks and valleys are relatively regular. If the valley area is represented by a resistance  $R_1$ , and the peak area resistance by  $R_2$ , as shown in Fig.4 (b), the resistances of a uniform and non-uniform finger can be described by (16) and (17), respectively.

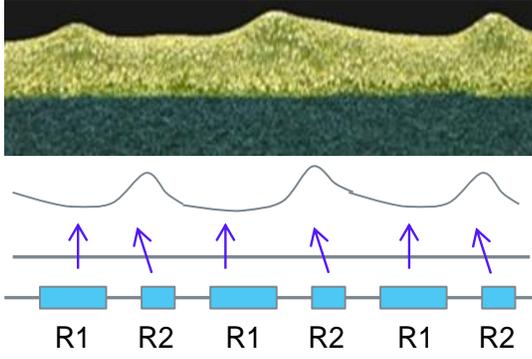


Fig.4 Resistance model for finger roughness

$$f = \frac{R_2}{R_1} \quad (15)$$

$$R_{uniform} = R_1 + R_1 + \dots = 2MR_1 \quad (16)$$

$$\begin{aligned} R_{non-uniform} &= R_1 + R_2 + R_1 + R_2 + \dots \\ &= M(1 + f)R_1 \end{aligned} \quad (17)$$

$$R_{non-uniform} = \frac{1+f}{2} R_{uniform} \quad (18)$$

If the ratio of peak and valley resistance is defined by roughness coefficient  $f=R_2/R_1$ , then the finger power loss expression (13) and (14) can be modified as (19) and (20), respectively.

$$P_{finger1} = \frac{4}{3} J_L^2 n a^3 b^2 \rho_f \frac{1+f}{(w - \frac{t_f}{tg\alpha}) t_f} \quad (19)$$

$$P_{finger2} = \frac{8}{3} J_L^2 n a^3 b^2 \rho_f \frac{\sqrt{\ln 2} (1+f)}{s w t \sqrt{\pi}} \quad (20)$$

In addition to emitter, base, fingers, the fourth contribution to power loss is the contact resistance. Considering the transfer length concept, the contact resistance [2] is given by,

$$R_c = \frac{L_T R_{sheet}}{a} \coth\left(\frac{W_f}{2L_T}\right) \quad (21)$$

Where  $L_T$  is the transfer length [3],

$$L_T = (\rho_c / R_{sheet})^{1/2} \quad (22)$$

In the case that  $W_f/2 > 1.5L_T$ ,  $\coth(W_f/(2L_T)) \cong 1$ , and  $R_c = \rho_c/(L_T a)$ ; so contact resistance is independent of grid width  $W_f$ . However, with the present trend to reduce finger width for multi-crystalline and mono-crystalline silicon solar cell ,

typically around 40~90  $\mu\text{m}$ , this limit is not valid. In hence, the finger with will influence the contact resistance value, as shown in equation (21).

Since the power loss associated with front contact resistance is  $2nI^2R_c$ , and  $I=J_Lab$ , so the expression of contact resistance dissipated power yields,

$$P_{contact} = 2J_L^2 nab^2 (\rho_c R_{sheet})^{1/2} \coth\left(\frac{W_f}{2} \sqrt{\frac{R_{sheet}}{\rho_c}}\right) \quad (22)$$

The back contact is covered by large area of metal (Al) and can be ignored [1] compared to other series segment.

For bus-bar, we can also use a trapezoid shape to describe the cross-section area. In the unit cell,  $W_b$  is half of bus-bar width (Fig. 5), so the bus-bar power loss is described as (23),

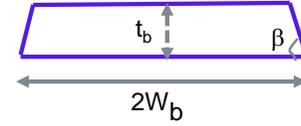


Fig. 5 Cross section area of bus-bar

$$P_{bus} = \frac{4}{3} J_L^2 n^3 a^2 b^3 \frac{\rho_f tg\beta}{(2tg\beta W_b - t_b) t_b} \quad (23)$$

Since the bus-bar width is much wider than that of finger, it is easier to print the bus-bar with a rectangular cross-sectional area, and the expression can be deduced to (24),

$$P_{bus} = \frac{2}{3} J_L^2 n^3 a^2 b^3 \frac{\rho_f}{W_b t_b} \quad (24)$$

If the energy conversion efficiency of solar cell is  $\eta$ , and the power density is  $P_L$ , the front metal shadowing caused power loss is

$$P_{shadow} = P_L \eta n [(a - w_b)w_f + 2bw_b] \quad (25)$$

The overall power loss is the sum of power loss from the emitter, contact, grid fingers, bus-bar, base, plus the shadowing loss,

$$P_{Rs} = P_{sheet} + P_{contact} + P_{finger} + P_{bus} + P_{base} \quad (26)$$

$$P_{all} = P_{Rs} + P_{shadow} \quad (27)$$

Equations (1), (2), (19), (20), (22), (23), and (25) are then normalized to unit cell area  $2nab$ , and the summary of expressions are listed in TABLE-I.

TABLE I  
SUMMARY OF POWER LOSS EXPRESSION

Power Loss Factors	Normalized expression
Emitter	$P_{sheet} = \frac{1}{3} J_L^2 b^2 R_{sheet}$
Front Contact	$P_{contact} = J_L^2 b (\rho_c R_{sheet})^{1/2} \coth\left(\frac{w_f}{2} \sqrt{\frac{R_{sheet}}{\rho_c}}\right)$
Grid finger	$P_{finger1} = \frac{2}{3} J_L^2 a^2 b \rho_f \frac{1+f}{(w - \frac{t_f}{tg\alpha}) t_f}$
Grid finger	$P_{finger2} = \frac{4}{3} J_L^2 a^2 b \rho_f \frac{\sqrt{\ln 2(1+f)}}{s w_f t_f \sqrt{\pi}}$ , $s = \frac{FWHM}{w_f}$
Bus-bar	$P_{bus} = \frac{2}{3} J_L^2 n^2 b^2 a \frac{\rho_f t g}{(2 t g \beta W_b - t_b) t_b}$
Base	$P_{base} = J_L^2 l \rho_{base}$
Shadowing	$P_L \eta \left[ \frac{(a - w_b) w_f}{2 a b} + \frac{w_b}{a} \right]$

The cell series resistance, normalized to unit area is,

$$r_s = \frac{P_{Rs}}{J_L^2} \quad (28)$$

The segment of each cell series resistance, normalized to unit area, can be achieved by substituting  $P_{Rs}$  with the corresponding power loss factor from TABLE I.

### III. RESULTS AND DISCUSSIONS

Two standard multi-crystalline silicon solar cells were printed using Heraeus 96XX front Ag paste with a 3 bus-bar 85-finger line screen, and the back metallization was formed by a commercial Al paste.

The related device and material parameters are listed in TABLE-II. Using these parameters, we can estimate the normalized power loss factors in TABLE-I and the corresponding cell series resistance.

TABLE-III shows the comparison of estimated series resistance values to the I-V tested values. When finger roughness coefficient  $f=1$ , i.e. grid fingers are uniform; the estimated series resistance using rectangular, trapezoid, and Gaussian cross-section area finger modes for cell-1 are 0.41, 0.44, and 0.48  $\Omega\text{-cm}^2$ , values much smaller than the tested value 0.66  $\Omega\text{-cm}^2$ . When the roughness effect is taken into consideration ( $f=3.25$ ), the calculated values increased to 0.50, 0.56, and 0.66  $\Omega\text{-cm}^2$ , correspondingly, and are much more closer to the tested value.

TABLE II  
CELL 1&2 DEVICE AND MATERIAL PARAMTERS

Parameters	Cell-1	Cell-2
$R_{sheet} (\Omega/\square)$	81.9	80.6
$t_f (\mu\text{m})$	17.2	17.4
$w_f (\mu\text{m})$	78.3	84.2
$t_b (\mu\text{m})$	15.5	15.7
$W_b (\text{mm})$	0.75	0.75
$\eta$	17.6	17.4
$a (\text{cm})$	2.55	
$b (\text{cm})$	0.091	
$\rho_c (\Omega\text{-cm}^2)$	$2.3 \times 10^{-3}$	
$\rho_f (\Omega\text{-cm})$	$2.7 \times 10^{-6}$	
$l (\mu\text{m})$	170	
$S=FWHM/w_f$	0.49	
$\alpha$	$40^\circ$	
$J_L (\text{mA/cm}^2)$	38.2	

TABLE III  
CALCULATED & TESTED SERIES RESISTANCE

Items	Cell-1		Cell-2	
	1	3.25	1	3.25
Rs-Rectangular ( $\Omega\text{-cm}^2$ )	0.41	0.50	0.40	0.48
Rs-Trapezoid ( $\Omega\text{-cm}^2$ )	0.44	0.56	0.42	0.53
Rs-Gaussian ( $\Omega\text{-cm}^2$ )	0.48	0.66	0.47	0.63
Rs- Tested ( $\Omega\text{-cm}^2$ )	0.66		0.58	

The relationship of series resistance and roughness coefficient  $f$  for three types of finger models are shown in Fig.6. The results suggest that finger uniformity is very important, because the series resistance value is sensitive to finger roughness, hence cell performance will be affected.

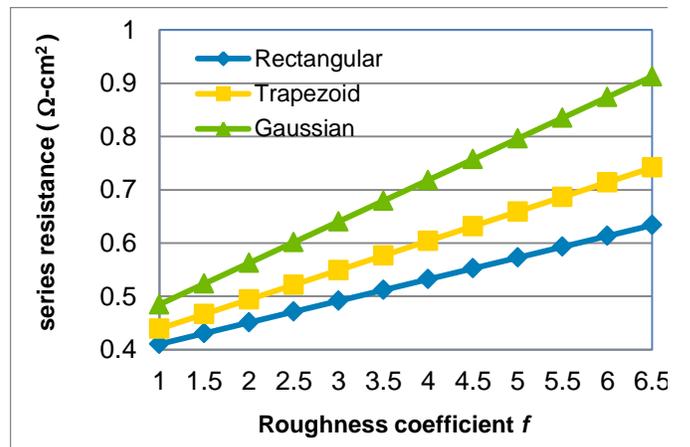


Fig.6 series resistance vs. roughness coefficient

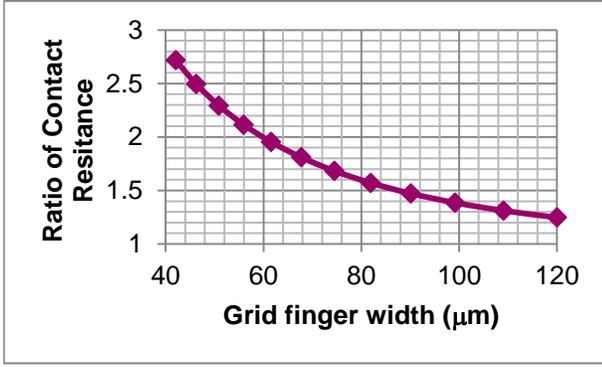


Fig. 7 Finger width influence on contact resistance

Fig. 7 presents the influence of grid finger width on contact resistance. The vertical axis is the ratio of contact resistance estimated using the revised contact model in TABLE-I and contact resistance without finger width influence, i.e.  $\coth(W_f/(2L_T)) \cong 1$ . The results show that when finger width is reduced, the contact resistance becomes larger. Therefore, its influence on cell performance will be more sensitive than wider fingers.

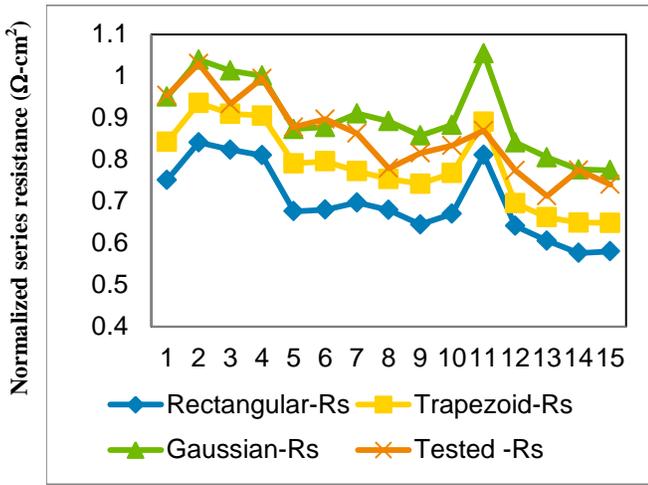


Fig. 8 Estimated and tested series resistance values of 15 samples

Finally, more samples (15 cells) were printed with various experimental pastes, and series resistance values are compared to the estimated ones. Fig.8 shows normalized cell series resistance value  $r_s$  of three finger type models and the I-V test results. From the chart, it is found that 8 samples showed best match using Gaussian finger model, 3 samples showed best

match using Trapezoid finger model, and 5 samples are between Gaussian and Trapezoid finger shape.

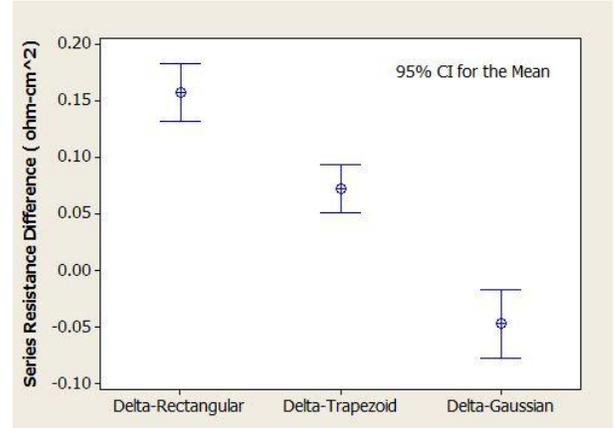


Fig. 9 Interval plot of series resistance difference between tested value and estimated value

The difference between the estimated results and the tested value is summarized in Fig. 9, where Delta-Rectangular, Delta-Trapezoid, and Delta Gaussian represents the difference of corresponding values to tested values. The accuracy, i.e. estimated value divided by tested value, was improved by 10%~20%.

#### IV SUMMARY

Three modifications are applied on conventional H-type grid pattern power loss model: 1) Gaussian or trapezoid cross-section area is used according to the practical screen printing finger shapes; 2) Non-uniformity effects are taken into consideration in the model; 3) The influence of finger width on contact resistivity was included. Real I-V test data is used to verify the estimated data, and the accuracy of the model is found to be improved.

#### REFERENCES

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